

blocks [E] and/or [F], wherein  $S_1, S_2, \dots, S_n$  are serial input bit lines of said architecture [A],  $n$  representing a number of coefficients of the filter transfer function;

wherein said architecture [A] includes  $m$  combinational logic blocks [B] consisting of full adder and full subtractor elements and the blocks [B] provide the addition terms of the filter transfer function  $[(a_0 * S_1 + b_0 * S_2 + \dots + k) * S_n], (A_1 * S_1 + b_1 * S_2 + \dots + k_1 * S_n) \dots (a_m * S_1 + b_m * S_2 + \dots + k_m * S_n)$ , where  $a_0, b_0 \dots k_0, a_1, b_1 \dots k_1, a_m, b_m, \dots k_m$  are (+/-1 or 0);

wherein the connection of the FA and/or FS elements to the  $S_1, S_2 \dots S_n$  lines and interconnection of the FA and FS elements depend on the value of coefficients  $a_0 \dots k_m$ ;

wherein the final output of a last FA or FS element of each block [B] is terminated through lines  $b_1, b_2 \dots b_m$  at a plurality of delay elements [T] within a logic block [C], the number of delay elements [T] depending on the size of a maximum coefficient value of the filter transfer function and being shareable by blocks [B];

wherein in said architecture [A] the blocks [B] are clustered together in series to form block [D] and delay elements [T] are clustered together in block [C], thereby spatially separating the sequential and combinational logic blocks;

wherein in said architecture [A], the output of each delay element [T] is connected to one of the inputs of one of the FA or FS elements of a respective block [B] corresponding to a next bit position;

wherein the interconnections between blocks [C] and [B] are represented as  $t_1, t_2, \dots, t_m$ , the FA and/or FS elements are arranged in matrix form such that elements  $FA_{0\_0}$  to  $FA_{0\_n}$  correspond to bit position 0, elements  $FA_{1\_1}$  to  $FA_{1\_n}$  correspond to bit position 1, and elements  $FA_{m\_1}$  to  $FA_{m\_n}$  correspond to bit position  $m$ ;

wherein the carry-out pin of each FA or FS element of each block [B] is fed to the input of an FA or FS element of a previous block [B] such that the same delay element [T] is used for multiplication by a factor of two and also for the carry structure in a one bit serial adder function;

wherein in the said architecture [A] logic components represented as block [Ex] are used for connecting the carryout of the FA and FS elements in a last stage of block [D] and a plurality of FA and/or FS elements and delay elements [T] are used within one block [Ex];

whereby said architecture [A] is structured into sequential block [C] consisting of delay elements and combinational block [D] consisting of FA and/or FS elements, while the delay elements [T] of block [C] are each positioned at a respective end position of each block [B], block [D] includes combinational element blocks [B] which essentially include FA and/or FS elements, thereby forming shareable logic elements within block [D]; and

a final output of the architecture [A] is a bit in the  $m^{\text{th}}$  position.

2. The device as claimed in claim 1 wherein when operated in bit serial fashion, the device provides hardware minimization of a finite impulse response filter, an infinite impulse response filter or for other filters, and application related to combinational logic consisting of delay elements [T], multiplier elements [M], and FA and/or FS elements.

3. The device as claimed in claim 1 wherein block [D] uses an FS element instead of an FA element when a filter transfer function coefficient value is close to a power of two.

4. The device as claimed in claim 1 wherein block [D] minimizes the use of FS elements by using a common subtraction operator and substituting FA elements instead.

5. (Amended) The device as claimed in claim 1 wherein coefficient lines CLin\_0... Clin\_n are not derived from a common input line but are instead respectively delayed by 0...n unit delays prior to input into said architecture [A].

6. (Amended) The device as claimed in claim 1 wherein instead of FA and FS elements, sequential adder and sequential subtractor elements are used.

7. A bit serial FIR filter device including:  
a logic block [A] adapted to receive an  $(m+1)$ -bit input and to produce a transfer function output corresponding to the  $m^{\text{th}}$  bit position, block [A] including:

a combinational-sequential logic block [D] adapted to receive a filter transfer function coefficients  $S_1, S_2, \dots S_n$  or a predetermined transfer function and including  $m+1$  combinational logic blocks  $B_0, B_1, \dots B_m$ ; and

a sequential logic block [C] having  $m$  delay elements  $T_1, T_2, \dots T_m$  for receiving respective outputs of blocks  $B_0, B_1, \dots B_{m-1}$  and for providing delayed outputs to respective blocks  $B_1, B_2, \dots B_m$ ;

wherein each block  $B_x$ , for  $x=0, 1, \dots, m$ , includes a plurality of serial subtractor or adder elements [SA], up to a maximum of  $n$ ,  $SA_{x\_1}, SA_{x\_2}, \dots SA_{x\_n}$ , for providing a coefficient multiplication function for each block  $B_x$ ; and

wherein block  $B_m$  outputs said transfer function output according to said transfer function, based on said  $(m+1)$ -bit input.

Please add new claims 8-17 to read as follows:

8. (New) A digital filter device comprising:

a coefficient circuit having  $n$  input bits corresponding to  $n$  filter coefficients;

a combinational circuit comprising a plurality of full adders with the interconnection of the plurality of full adders depending on values of the  $n$  coefficients to implement the addition terms of the filter transfer function; and

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a sequential circuit comprising a predetermined number of unit delays with the predetermined number being dependent on a maximum value of the  $n$  coefficients, the inputs of the unit delays being coupled to output of selected ones of the full adders and the output of the unit delays being coupled to an input of a full adder in an adjacent bit position from the corresponding unit delay.

9. (New) The filter of claim 8 wherein the combinational circuit implements the addition terms of the filter transfer function using the following form:

$(a_0 * S_1 + b_0 * S_2 + \dots + k) * S_n$ ,

$(a_1 * S_1 + b_1 * S_2 + \dots + k_1 * S_n)$ ,

...

$(a_m * S_1 + b_m * S_2 + \dots + k_m * S_n)$ ],

where  $S_1, S_1, S_n$  are filter coefficients and  
 $a_0, b_0 \dots k_0, a_1, b_1 \dots k_1, a_m, b_m, \dots k_m$  are (+/-1 or 0).

10. (New) The filter of claim 8 wherein the coefficient circuit receives  $n$  serial input bits.

11. (New) The filter of claim 8 wherein the combinational circuit and the sequential circuit are interconnected to implement a finite input response (FIR) filter.

12. (New) The filter of claim 8 wherein the combinational circuit and the sequential circuit are interconnected to implement an infinite input response (IIR) filter.

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13. (New) A method of implementing a digital filter comprising:  
generating  $n$  filter coefficients using a combinational logic circuit comprising a plurality of full adders with the interconnection of the plurality of full adders depending on values of the  $n$  coefficients to implement addition terms of the filter transfer function; and  
generating a plurality of multiplication factors sequential logic circuit comprising a predetermined number of unit delays with the predetermined number being dependent on a maximum value of the  $n$  coefficients, the inputs of the unit delays being coupled to output of selected ones of the full adders and the output of the unit delays being coupled to an input of a full adder in an adjacent bit position from the corresponding unit delay.

14. (New) The method of claim 13 wherein the digital filter has  $n$  serial input bits corresponding to  $n$  filter coefficients.

15. (New) The method of claim 14 wherein the  $n$  input bits are serial input bits.

16. (New) The method of claim 13 wherein the combinational circuit and the sequential circuit are interconnected to implement a finite input response (FIR) filter.